

and covariance V_k .

$$(3) \quad E(v_i v_j^T) = V_k$$

where V_k is a known symmetric nonnegative definite matrix.

It is assumed that x_1 , the initial state of the system, is a random vector from a distribution with known mean \bar{x}_{10} , the estimate of x_1 at time zero, and covariance V_1 .

$$(4) \quad E(x_1) = \bar{x}_{10}, \quad E(v_i v_j^T) = V_k, \quad v_i^T = x_i - \bar{x}_{10}.$$

Note that this can be written as the equation

$$(5) \quad x_{10} = x_1 - v_1^T$$

where v_1^T is a zero mean noise vector with known covariance V_1 .

The problem arises because the state vector x_k is not directly measurable; instead the m dimensional vector of observations y_k is available where it is known that for $k = 1, 2, \dots$,

$$(6) \quad y_k^T = C_k^T x_k + w_k^T$$

with C_k a known matrix and w_k^T a noise vector such that

$$(7) \quad E(w_k w_k^T) = 0, \quad E(w_k w_j^T) = W_k,$$

where W_k is a known nonnegative definite matrix.

Given all the information in (1)-(7) with the observations y_k^T available up to time k , we might want to find an estimate of x_k which is best in some sense. This is called *filtering* in the engineering literature (e.g. [12]). We might also want a best estimate of x_{k+j} for $j > 0$ this is *prediction*, while for $j < 0$ it is *smoothing*.

3. Weighted linear least squares formulation of the problem. Here all the information on the system and measurements up to time k will be given equal weight, and combined together to form one large equation. This can be regarded as a large overdetermined linear system with zero mean unit covariance noise, which can then be solved using linear least squares techniques.

From now on we will assume that the V_k and W_k are all positive definite so that we can transcend equations (1), (5) and (6) into equations with zero mean noise vectors having known covariance matrices, as a result we will be able to assume that all equations are equally important. In (3), (4), and (7) let

$$(8) \quad L_k^T x_k = V_k^{-1} x_k, \quad L_k^T w_k = W_k^{-1} w_k$$

be the Cholesky decompositions of the inverses of the covariance matrices. If these V_k and W_k are available then there is no difficulty. If a covariance matrix V is known, then its inverse, then the reverse Cholesky decomposition $R R^T = V$, R upper triangular, can be found (see, for example, [11, p. 125]). It then follows that $L = R^{-T}$ is the Cholesky factor in (8). The reverse decomposition of V has been chosen here as it leads to computational savings later when the method exhibited in (25) is used to triangulate the matrix in (24).

The factor L_k in (8) can be applied to (5) to give

$$d_k = L_k^T x_{10} = L_k^T x_1 - L_k^T v_1^T = L_k^T x_1 + v_1, \\ (9) \quad E(v_1) = 0, \quad E(v_1 v_1^T) = L_k^T V_1 L_k = I$$

Similarly if we write

$$(10) \quad F_{k-1} = -L_{k-1}^T F_{k-1}, \quad d_k = L_k^T B_k u_k, \quad v_k = -L_k^T v_k^T$$

then (1) may be rewritten as

$$(11) \quad d_k = [F_{k-1}, L_k^T] \begin{bmatrix} x_{k-1} \\ v_k \end{bmatrix} + v_k, \quad k = 2, 3, \dots,$$

with v_k representing a zero mean unit covariance noise vector.

With the same approach the observation equation (6) becomes

$$(12) \quad y_k = L_k^T y_k = L_k^T C_k x_k + L_k^T w_k = C_k x_k + w_k, \quad k = 1, 2, \dots,$$

where w_k is zero mean unit covariance noise.

Equations (9), (11), and (12) combine to give the formulation of the problem suggested by Duncan and Horn [3]

$$(13) \quad \begin{bmatrix} d_1 \\ d_2 \\ \vdots \\ d_k \\ y_1 \\ y_2 \\ \vdots \\ y_k \end{bmatrix} = \begin{bmatrix} L_1^T \\ C_1 \\ \vdots \\ F_{k-1} & L_k^T \\ \vdots & C_k \\ \vdots & \vdots \\ \vdots & \vdots \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_k \\ w_1 \\ w_2 \\ \vdots \\ w_k \end{bmatrix} + \begin{bmatrix} v_1 \\ v_2 \\ \vdots \\ v_k \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$

Equation (13) can be written more briefly as

$$(14) \quad y^T = F^T x + v^T,$$

though in the future we will usually drop the superscripts. We see that if for $i \neq j$

$$(15) \quad E(v_i v_j^T) = 0, \quad E(w_i w_j^T) = 0, \quad E(v_i w_j^T) = 0,$$

then $v^{(i)}$ is a zero mean unit covariance noise vector. Equation (15) will hold if and only if it holds with v and w replaced by v and w' respectively. For $i = j$ we assume (15) holds, this is the assumption to guarantee that $v^{(i)}$ is a unit covariance noise vector. It is possible to assume that $v^{(i)}$ and $w^{(i)}$ are independent, but this is not necessary. Clearly the last assumption could be modified to handle correlated noise.

The least squares solution to (14) is

$$(16) \quad \hat{x} = \arg \min \|F x - y\|^2.$$

that is, \hat{x} is the argument that minimizes the sum of squared residuals.

Rosenbrock [13] has observed that Kalman filtering is an extension of the work of Gauss. By putting the problem in the form of (13) and (14), Duncan and Horn have shown that it has *exactly* the same form as the problem that Gauss posed and solved.

4. **Numerical solution of the least squares problem.** In order to find the solution to (16) we use an orthogonal matrix Q such that

$$(17) \quad Q^T F = \begin{bmatrix} Q_1^T \\ 0 \end{bmatrix}, \quad F = \begin{bmatrix} R \\ 0 \end{bmatrix},$$

where R is upper triangular. Since the 2-norm is unaffected by orthogonal transformations,

$$(18) \quad \hat{x} = \arg \min \left\| \begin{bmatrix} R \\ 0 \end{bmatrix} x - \begin{bmatrix} Q_1^T \\ 0 \end{bmatrix} y \right\|,$$

from which it follows that \hat{x} satisfies

$$(19) \quad R\hat{x} = Q_1^T y.$$

Here R will certainly be nonsingular if the L_j 's are nonsingular in (13). Lawson and Hanson [11] treated the case when F was suggested by Golub [6]. Lawson and Hanson [11] treated the case when F has band form. Gentleman [5] showed how fast, numerically stable variants of the Givens rotation matrices without square roots could be used to gain speed in special cases like the present one. Here advantages will be taken of these techniques, and the special form of F , to obtain a fast recursive set of operations for its reduction to block upper bidiagonal form. It will be shown that (19) becomes

$$(20) \quad \begin{bmatrix} R_1 & R_{1,2} & & \\ R_2 & R_{2,3} & & \\ & & \ddots & \\ & & & R_{k-1} & R_{k-1,k} \\ & & & & \bar{R}_k \end{bmatrix} \begin{bmatrix} x_{1k} \\ x_{2k} \\ \vdots \\ x_{k-1,k} \\ x_{k,k} \end{bmatrix} = \begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_{k-1} \\ b_k \end{bmatrix},$$

where x_{jk} means the weighted linear least squares estimate of x_j , the state vector at time t_k given y_1, \dots, y_k , the observations up to time t_k . Thus $x_{kk} = x_k$, $j = 1, \dots, k-1$, where values x_{jk} is a filtered value, while $x_{k,k}$ would be a predicted value. The filtered values are as many smoothed values as are needed, can be found directly from (20) by back substitution.

The last block of the matrix \bar{R} in (20) is \bar{R}_k , rather than R_k , since it is the only one of these blocks that will be altered when the decomposition is updated to account for the introduction of new rows and columns, corresponding to new observations like (6), or new system equations like (1).

The matrix F in (13) and (14) can be transformed to the matrix R in (17) and (20) in a sequence of $2k-1$ steps. First we set

$$(21) \quad \hat{R}_1 = L_1^T;$$

then in steps $2j-1$ and $2j$ we develop an orthogonal matrix Q_j which transforms

$$(22) \quad \begin{bmatrix} \hat{R}_j & 0 \\ C_j^T & 0 \\ F_j^T & L_{j-1}^T \end{bmatrix} \text{ to } \begin{bmatrix} \hat{R}_j & R_{j,j+1} \\ 0 & 0 \\ 0 & \hat{R}_{j+1} \end{bmatrix}$$

so that when the zero rows are permuted to the bottom of the right hand matrix, it becomes upper triangular.

In step $2j-1$ an orthogonal matrix \hat{Q}_j is chosen to zero out C_j as follows:

$$(23) \quad \hat{Q}_j^T \begin{bmatrix} \hat{R}_j \\ C_j^T \end{bmatrix} = \begin{bmatrix} \hat{R}_j \\ 0 \end{bmatrix}$$

where \hat{R}_j is upper triangular and the orthogonal matrix \hat{Q}_j is the product of n Householder matrices chosen to take advantage of \hat{R}_j being upper triangular. Thus for $i = 1, 2, \dots, n$, the i th elementary matrix H_i of \hat{Q}_j is chosen to zero out the altered C_j matrix to the right of the element \hat{R}_{ji} . The complete step will take approximately $2n^2$ multiplications. Equation (20) shows the complete system approximation of this form has just been completed. Next, in step $2j$, the orthogonal matrix \hat{Q}_j is chosen to zero out F_j :

$$(24) \quad \hat{Q}_j^T \begin{bmatrix} \hat{R}_j & 0 \\ F_j^T & L_{j-1}^T \end{bmatrix} = \begin{bmatrix} \hat{R}_j & R_{j,j+1} \\ 0 & 0 \\ 0 & \hat{R}_{j+1} \end{bmatrix},$$

where full advantage of the structure can be taken by using a series of rotations as follows. For $i = 1, \dots, 1$, the i th row of F can be set to zero an element at a time by using the i th row of the altered \hat{R}_j to eliminate the i th element of the i th row of the altered F ; this must be done in the order $i = 1, 2, \dots, n$. The start of this step can be indicated schematically in the case of $n = 3$ by

$$(25) \quad \begin{array}{ccccccc} x & x & x & & & & \\ \downarrow & & & & & & \\ x & x & x & & & & \\ \downarrow & & & & & & \\ x & x & x & x & x & & \\ \downarrow & & & & & & \\ x & x & x & x & x & x & x \end{array} \quad \begin{array}{c} \textcircled{1} \\ \textcircled{2} \\ \textcircled{3} \\ \textcircled{4} \\ \textcircled{5} \end{array}$$

where $\textcircled{1}$ indicates this element is set to zero in the i th rotation by combining it with the element at the other end of the arrow, and $\textcircled{2}$ indicates this originally zero element has been made nonzero in the i th rotation. Other orders of elimination are possible, but this appears to be the quickest, taking about $4n^2$ multiplications using ordinary rotations. If n is large enough to justify the overhead involved in using stable two-multiplication rotations [5], [7], this can be reduced to $2n^2$. More will be said on this in § 8.

and we see from (35) that rows $(l-1)m+1$ to m of $R^{(l-1)}$ are

$$[0, \dots, 0, \hat{R}_l^T, 0, \dots, 0]$$

and as a result

$$(37) \quad H_{kl} = \hat{R}_l^T \hat{R}_l^T = (\hat{R}_l^T \hat{R}_l)^{-1}.$$

We see then that factors \hat{R}_l of the inverses of as many error covariance matrices as are wanted may be computed simply by carrying out the bidiagonalization in (35) as far as we want, to produce a sequence $\hat{R}_k = R_k$ and $\hat{R}_j^T = R_j^T$, $j = k-1, k-2, \dots$.

Probably the best way of computing each step of this block bidiagonalization in R is to use the reverse of the method (25) that was used for computing (24); in this way full account is taken of sparsity (25) and the resulting \hat{R}_k are upper triangular in (37), so these will be the Cholesky factors of the covariance matrix inverses. The next step in (35) can be described as

$$(38) \quad P^{r-1} = \begin{bmatrix} I_{m-r+1} \\ P^{r-1} \end{bmatrix}, \quad P^{r-1} \begin{bmatrix} R_{r-1} & 0 \\ R_{r-1} & \hat{R}_r \end{bmatrix} = \begin{bmatrix} \hat{R}_{r-1} & 0 \\ \hat{R}_{r-1} & \hat{R}_r \end{bmatrix}$$

where, for $r=1, 2, \dots, m$, the l th column of R_{r-1} is set to zero by a series of rotations of the l th row of \hat{R}_r with the s th row of the altered R_{r-1} , $s = m, m-1, \dots, l$. The start of this can be described as

$$(39) \quad \begin{array}{ccccccc} x & x & x & \textcircled{1} & x & x & x \\ & x & x & x & \textcircled{2} & x & x \\ & & x & x & x & \textcircled{3} & x \\ & & & x & x & x & \textcircled{4} \\ & & & & x & x & x \\ & & & & & x & x \\ & & & & & & x \end{array}$$

where the same notation is used as was used in (25). Again using two-multiplication rotations this takes $2m^2$ multiplications.

This method is consistent with the present approach of computing factors of covariance matrices directly, and its accuracy is an attractive feature. However it is not clear how the results could be updated from step to step, and so if several of these covariance matrices are wanted regularly, this would be expensive. Another approach is to define the $k \times m$ matrix N_k having the unit matrix in rows $(l-1)m+1$ to m , and zero elsewhere. Then

$$H_{kl} = N_k^T (R^T R)^{-1} N_k = S_k^T S_k = R_k^T R_{kl}$$

where $R_k^T S_k = N_k$, and the Cholesky factor R_{kl} of H_{kl} could be found by using orthogonal rotations in triangularize S_k . A little thought shows such factors can easily be updated when new information arrives. This appears to be a fast and workable method, but the only possible inaccuracies arising in the solution of equations for S_k if these are ill conditioned. However this method gives us factors of the covariance matrices, in contrast to the rest of the paper which deals with

factors of inverses of covariance matrices. For reasons of space and consistency we will not pursue it further.

6. Further predicted values and their covariances. If any further values of the state vector need be predicted on the basis of the measurements up to and including time k , then these follow directly from (1)

$$(40) \quad x_{k+j} = F_{k+j-1} x_{k+j-1} + B_{k+j} u_{k+j}$$

for $j=2, 3, \dots$. These can be thought of as the best estimates for the model

$$(41) \quad \begin{bmatrix} y^{(k)} \\ d_{k+1} \\ d_{k+j} \end{bmatrix} = \begin{bmatrix} F^{(k)} & & \\ O & L_{k+1}^T & \\ & & L_{k+j}^T \end{bmatrix} \begin{bmatrix} x^{(k)} \\ x_{k+1} \\ x_{k+j} \end{bmatrix} + \begin{bmatrix} g^{(k)} \\ w_{k+1} \\ w_{k+j} \end{bmatrix}$$

which is the continuation of (14) if no further measurements are available. Here the zero mean random noise vector has unit covariance, and so the upper triangular Cholesky factor of the inverse of the covariance matrix of the error $\hat{x}_{k+j} = x_{k+j} - \hat{x}_{k+j}$ is just the bottom right hand corner matrix obtained when the upper triangularization of $F^{(k)}$ in (17) is continued to the new matrix in (41). This can be done using j steps of the same form as was used for (24).

7. An extension of the model. Although in equations (1) and (6) the state at time $k+1$ and the observation at time k are given in terms of the state at time k only, the model can be more general than this, as is shown for example by Hannan [8]. One possibility is that x_{k+1} can depend linearly on $x_k, x_{k-1}, \dots, x_{k-j}$ and y_k can depend linearly on $x_k, x_{k-1}, \dots, x_{k-j}$. The present formulation and computational solution can easily be extended to such cases, for example with $j=f-1$, the equivalent of (13) for $k-3$ might be

$$\begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = \begin{bmatrix} L_{f-1}^T & C_{1,1} & L_f^T \\ & C_{1,1} & L_f^T \\ & C_{2,1} & C_{2,2} \\ & & F_{2,1} & L_{f-1}^T \\ & & & F_{2,1} & L_{f-1}^T \\ & & & & C_{2,2} & C_{2,3} \end{bmatrix} \begin{bmatrix} x_1 \\ w_1 \\ x_2 \\ w_2 \\ x_3 \\ w_3 \end{bmatrix}$$

Here the first three blocks of rows would be transformed to upper triangular form just as in (22); then this complete upper triangle would be used to eliminate $C_{2,1}$ and $C_{2,2}$ as in (23), and the resulting upper triangle would be used to eliminate $F_{2,1}$ and $F_{2,2}$ and transform L_{f-1}^T as in (24). The lower part of the resulting upper triangle would then be used to eliminate $C_{2,3}$ and $C_{3,1}$, and so on.

8. Operation count. A simple generalization of the transformations described by (24), (25) and (38), (39) can be used very effectively in other

computations in estimation, as well as in optimal control calculations. The generalization of (24) can be simply described as

$$(42) \quad Q^T \begin{pmatrix} R_1 & 0 \\ G & R_2 \end{pmatrix} = \begin{pmatrix} \tilde{R}_1 & \tilde{R}_{12} \\ 0 & \tilde{R}_2 \end{pmatrix}$$

where O is orthogonal and R_1 and \tilde{R}_1 are $n \times n$ upper triangular matrices, but now R_2 and \tilde{R}_2 are $m \times m$ upper triangular. It is worthwhile considering this slightly more general case in order to compare the use of rotations, as described in §4, with reduction by Householder transformations.

Using the Householder transformations fills up the block R_2 in the first step, and requires $m(2m+n)$ multiplications to produce upper triangular \tilde{R}_1 , leaving \tilde{R}_2 full. It then takes another $2m^2$ multiplications to make \tilde{R}_2 upper triangular, a total of

$$(43) \quad mn(2m+n) + 2m^2/3$$

multiplications for the complete Householder reduction. Standard rotations require 4 multiplications each, while it is possible to compute stable two- or three-multiplication rotations [5], [7] with some added overhead. If we use k -multiplication rotations, then the complete reduction can be carried out in

$$(44) \quad kmn(m+n) + 2m^2/3$$

multiplications. These results show that four-multiplication rotations are faster than Householder transformations when $m > 1.25n$, but in any case never take more than twice as long. In the present case with $m = n$ the comparison is $11n^2/3$ against $4n^2$, so Householder transformations have a (negligible) 8% advantage. An advantage of the technique described in §4 is that it requires no extra storage, since only one nonzero element is generated as the zero's product. In contrast, Householder transformations immediately introduce $m^2/2$ nonzero elements that later have to be made zero. For this reason we recommend using rotations at all times: ordinary rotations are satisfactory if there is no time problem, while stable two-multiplication rotations can be used if the dimensions of the problem are large enough to warrant them, and these are superior to Householder transformations for any ratio of m and n .

For the complete operation count here, we assume that two multiplication rotations have been used for (24), taking $2n^2$ multiplications. Now if we assume V_1^T and W_1^T are supplied in (8), then computing the Cholesky decompositions and forming $L_1^T P_1^{-1}$ in (10) and $L_1^T C_1$ in (12) takes $n^2/6 + m^2/6 + n^2/2 + mn^2/2$ multiplications. Computing (23) takes mn^2 multiplications, to give a total for one time step of

$$(45) \quad 8n^3/3 + m^2/6 + mn^2 + mn^2/2.$$

This provides the Cholesky factors of the inverses of H_0 and H_{10} . The time required to compute the right hand side in (24) and solve for the estimates x_{10} is small in comparison to the other computations.

9. **Comparison with other algorithms.** In order to compare algorithms it will be useful to discuss briefly some of the important developments since Kalman's original approach. A more complete discussion is given in [10]. The major computation in Kalman's method was to update the covariance matrix $H_{k|k-1}$ in order to produce $H_{k|k}$ and $H_{k+1|k}$ as follows.

$$(46) \quad \begin{aligned} H_{k|k} &= H_{k|k-1} - H_{k|k-1} C_1^T [C_1 H_{k|k-1} C_1^T + C_1^T + V_k]^{-1} C_1 H_{k|k-1} \\ H_{k+1|k} &= F_1^T H_{k|k} F_1^T + V_k \end{aligned}$$

where we have used the notation in (1), (3), (6), and (7). Matrices which are represented by symmetric capitals here are nonnegative definite in theory, but in computations this property was often lost. Later workers avoided loss of definiteness by updating square roots of matrices (where S was said to be a square root of the matrix M if $SM = S^T S$). This also avoided inaccuracies caused by squaring, since it was possible to use CS, the square root of CHC^T , rather than CHC^T itself. The covariance matrices are given in (3) and (7) for the *constant square root filter*. If covariance matrices are given in (3) and (7) for the *noise*, and in (4) for the initial estimate, and if covariance matrices of the estimates are desired as output, then covariance square root filtering appears to be the correct approach.

The inverse of a covariance matrix is called an *information matrix*. It is possible to update information matrices for this problem, but the same arguments on definiteness and accuracy of square roots apply here too, and so it is numerically superior to consider *information square root filters*. The present method requires only factors of information matrices in (8), and gives factors of information matrices in §5, so it is clearly an information square root filter.

One of the best available approaches for this estimation problem was given by Dyer and McReynolds [4], and this has also been called an information square root filter. They considered zero mean unit covariance noise as in (11) and (12), and sought to minimize

$$(47) \quad J(k) = \|L_1^T(x_1 - x_1^0)\|^2 + \frac{1}{2} \|e_1\|^2 + \frac{1}{2} \|e_2\|^2$$

subject to the constraints (11) and (12). In order to estimate x_1 , an implementation which is as numerically stable as this particular approach allows has been described by Hanson and Dyer [14], and is in use as a navigational tool at the Jet Propulsion Laboratory in Pasadena, California.

Although apparently the most squares-root formulation of the problem, Dyer and McReynolds apparently did not consider it in the form of the *constant square matrix* problem in (13), and perhaps for this reason they required the *information* transformation (4, (26))

$$(48) \quad P \begin{bmatrix} I & 0 \\ R_1(R_1^T L_1^T L_1 + R_1(R_1^T))^{-1} R_1(R_1^T) \end{bmatrix} = \begin{bmatrix} N & B \\ 0 & R_{k+1} \end{bmatrix}$$

where we have used the terminology of (1), (3), (8), and (12) to (14). The equivalent transformation here is given by (24). The difficulty is that (48) requires the equivalence of the computation of the inverses of V_k^T , and so cannot be carried

out when F_1 is singular, and leads to numerical inaccuracies in general unless F_1 is well-conditioned. Problems with poorly conditioned F_1 are not at all rare. Such difficulties do not occur in the present algorithm using the computation in (24), nor in well designed covariance square root filters, and so this appears to be a serious drawback of the Dyer and McReynolds information square root filter and its implementation in [14].

In one sense the Dyer and McReynolds method is halfway between a covariance square root filter and an information square root filter such as the one here, for although Dyer and McReynolds produce factors of information matrices, and require the information matrix of the measurement noise in (6), they require the covariance matrix of the system noise in (1). This suggests the possibility of a complete range of filters extending from the all covariance square root filter to the all information square root filter as outlined here.

An operation count of the Dyer and McReynolds method as described in [4] shows that it takes about

$$(49) \quad m^2/6 + m^2n/2 + mn^2 + 16n^2/3$$

multiplications for the problem described here if the covariance matrix of the system noise and the information matrix of the measurement noise are given. This is $8n^2/3$ more multiplications than in (43), which if $m = n$ means it takes about 60% more multiplications than the method given here. There are other variants of the Dyer and McReynolds algorithm as described in [1] and [10] which give computation counts more comparable with (43) but they also suffer from requiring the inverse of F_1 .

10. Comments. If some observations are noise free the corresponding covariance matrices will be singular. In such cases the method given here will not work (although it might be extended to allow for covariance matrices that are singular). Similarly Kalman's original method and other covariance updating methods have difficulties with singular or ill-conditioned information matrices. Thus the different basic methods have complementary applications.

The fact that information filters have complementary applications. The fact that information filters, like the one given here, can work in the absence of certain information, that is with singular information matrices, is an important advantage for some cases. Lack of certain information can cause leading submatrices of R in (20) to be singular; this can be detected in the present algorithm, and in such cases no estimates would be computed. But as more information comes in, nonsingularity of the later submatrices would allow estimates to be computed.

Again we emphasize that the way of formulating the problem suggested by Duncan and Horn is a natural and easily understandable approach and leads directly to good computational techniques. The well known fact that this weighted least squares result gives the best linear unbiased estimate for the case of nonsingular, bounded, noise covariance matrices; ties the result to Kalman's work in that it has to show the equivalence algebraically. The speed of the algorithm given here, together with its good numerical properties, makes it an attractive one when compared with other published algorithms.

Acknowledgments. We are very grateful for the support of the Department of Scientific and Industrial Research, Wellington where the initial part of this work was done, and for the hospitality of the people working there, especially Gordon Dyer. The interest lead to our reading the paper by Duncan and Horn. We also thank R. G. Anderson for programming the algorithm, and Mike Arbib, Gerald Bierman, Richard Hagan, and George Szyan whose comments helped our understanding of the subject.

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